

Laplace Transform summary

A. Tzanakopoulou, Prof. Ioannis Pitas
Aristotle University of Thessaloniki
pitass@csd.auth.gr
www.aiia.csd.auth.gr
Version 2.0.1

Laplace Transform

- **Definition**
- **Properties**
- **Inverse Laplace Transform**
- **Analysis of LTI Systems**

Definition

Unilateral Laplace transform (LT) is defined by:

$$X(s) = \int_0^{\infty} x(t)e^{-st} dt.$$

Bilateral Laplace transform definition:

$$X(s) = \int_{-\infty}^{\infty} x(t)e^{-st} dt.$$

As $s = \sigma + i\Omega$ is a complex number, LT is a mapping $\mathbb{R} \rightarrow \mathbb{C}$ of a signal on the complex plane.

Poles and Zeros

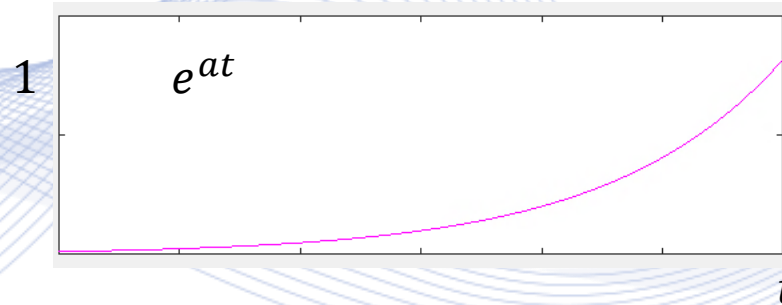
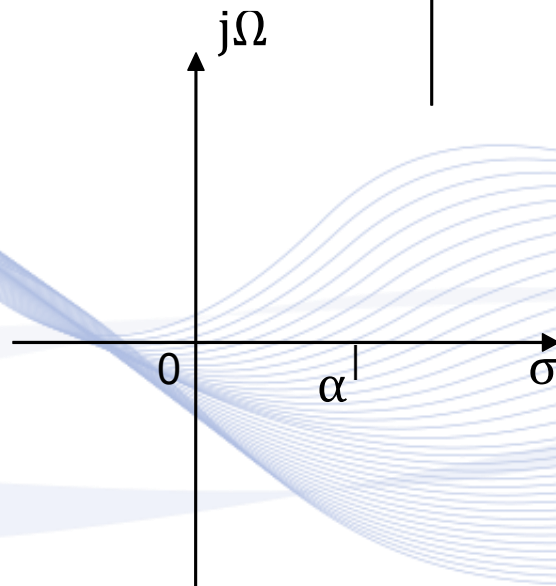
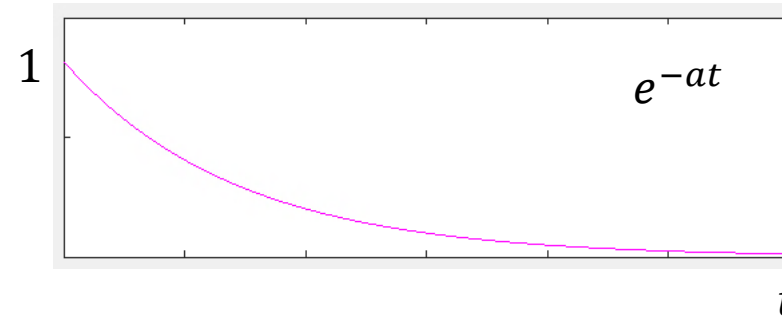
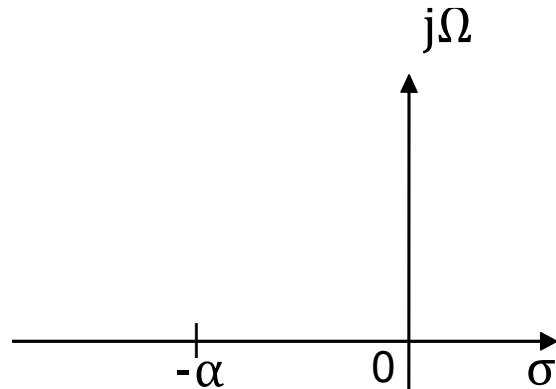
- If LT $X(s)$ is a rational function:

$$X(s) = \frac{N(s)}{D(s)},$$

the roots of $D(s)$ are named **poles** and the roots of $N(s)$ are named **zeros**.

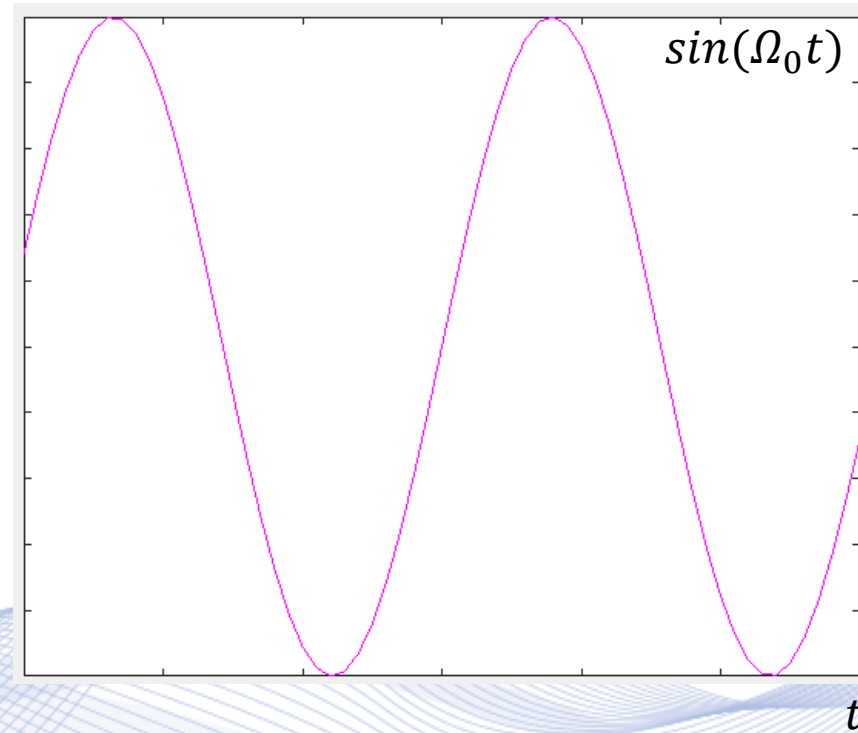
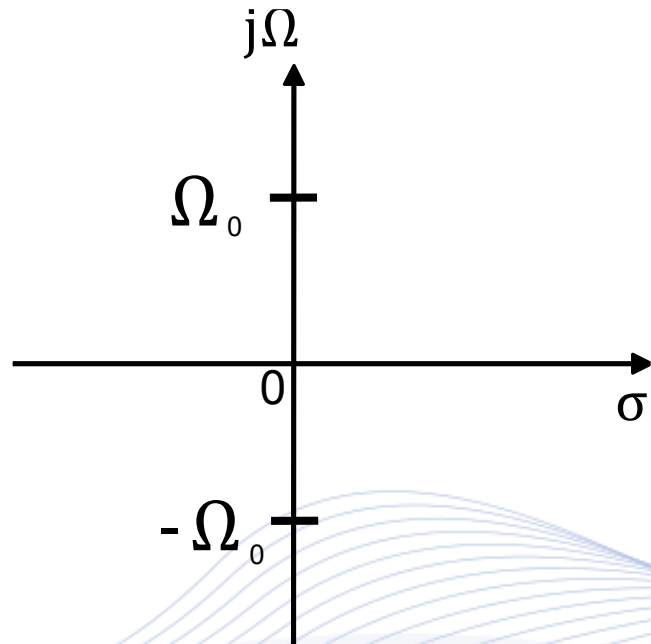
- Poles and zeros can be real or pair of complex conjugate numbers.

Poles and Zeros



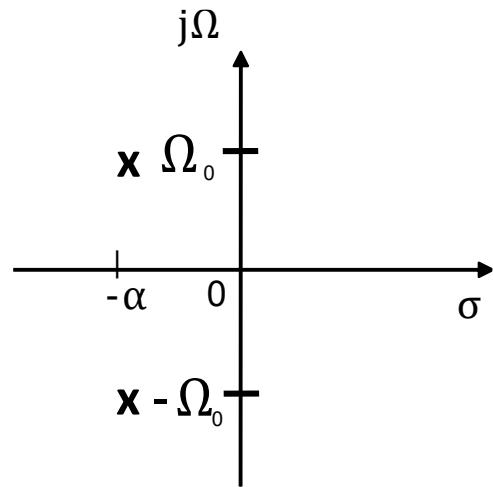
Single poles on the real axis.

Poles and Zeros

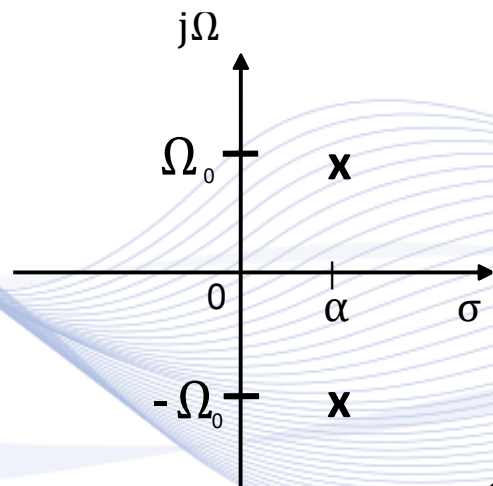
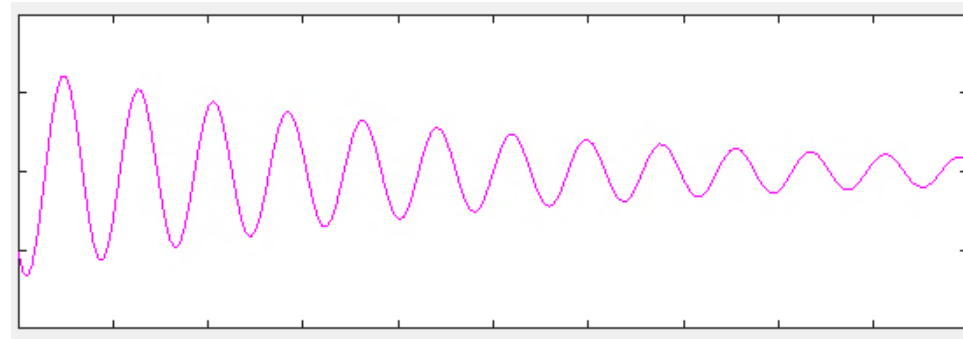


Multiple poles on the imaginary axis.

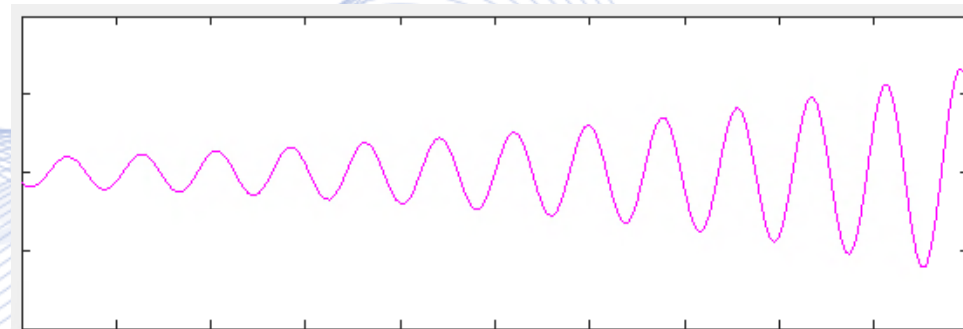
Poles and Zeros



$$e^{-at} \sin \Omega_0 t$$



$$e^{at} \sin \Omega_0 t$$

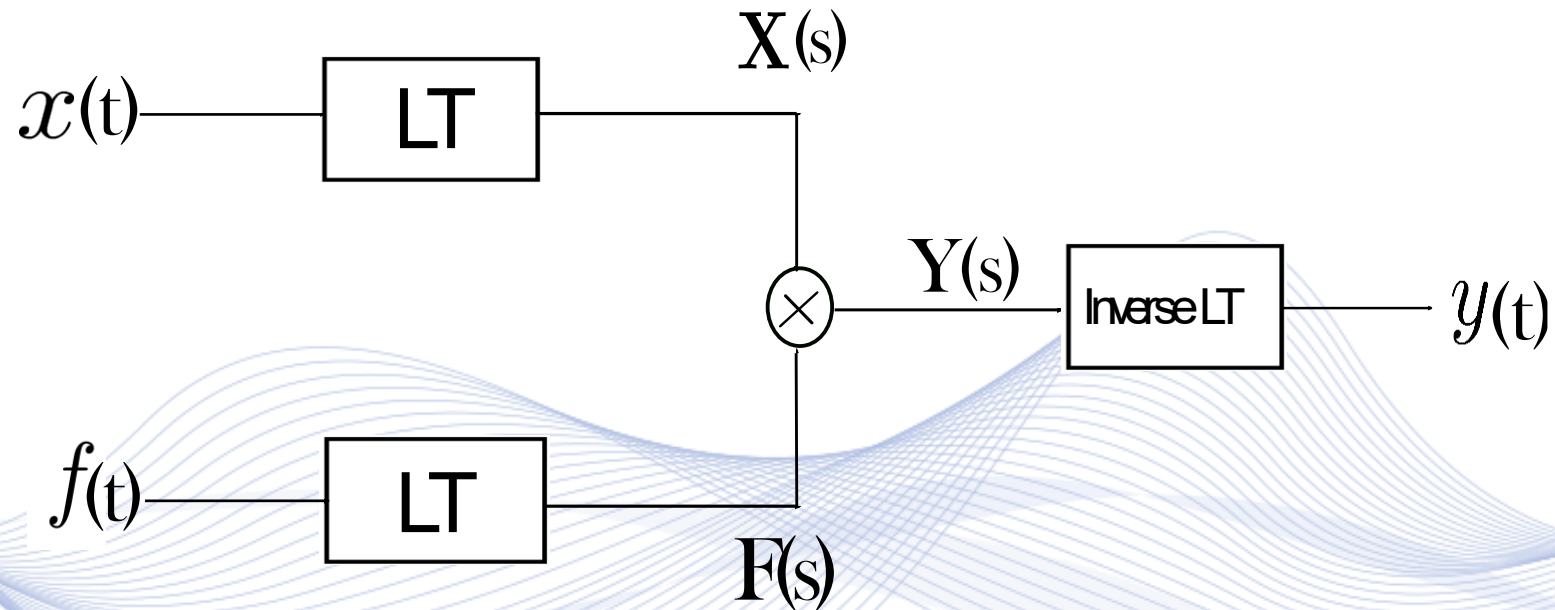


Complex conjugate poles.

Laplace Transform

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Laplace Transform Properties



Laplace Transform

- Definition
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- **Inverse Laplace Transform**
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Inverse Laplace Transform

The inverse LT is given by the ***complex inversion integral***:

$$x(t) = \frac{1}{2\pi i} \int_{\sigma - i\infty}^{\sigma + i\infty} X(s) e^{st} ds, \quad t > 0.$$

- Parameter σ should lie to the right of all singularities (poles), i.e., $\sigma > \sigma_0$.

Laplace Transform

- Definition
- Properties
- Inverse Laplace Transform calculation
- **Analysis of LTI Systems**

Transfer Function

An LTI system can be described by the convolution of its input $x(t)$ and its **impulse response** $h(t), t > 0$:

$$y(t) = \int_0^{\infty} h(\tau)x(t - \tau)d\tau.$$

By applying the convolution property of LT, we have:

$$Y(s) = H(s)X(s).$$

System **transfer function** $H(s)$ has a rational form:

$$H(s) = \frac{Y(s)}{X(s)}.$$

Transfer Function

If all the initial conditions are equal to zero, then by applying the LT in both parts of the previous formula, we have:

$$H(s) = \frac{Y(s)}{X(s)} = \frac{b_m s^m + b_{m-1} s^{m-1} + \dots + b_0}{a_n s^n + a_{n-1} s^{n-1} + \dots + a_0} = \frac{N(s)}{D(s)}$$

- $N(s), D(s)$ polynomials roots are called **zeroes** and **poles**, respectively.
- As, under certain conditions, $D(s)$ may become zero, leading to system instability, $D(s)$ defines system stability.

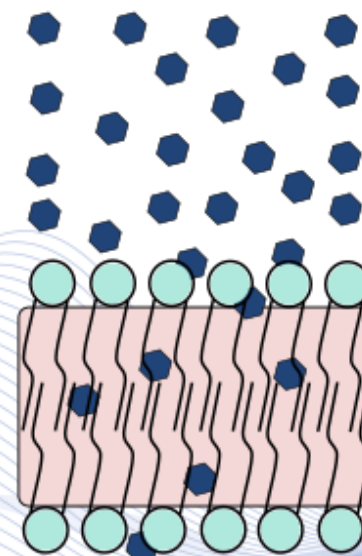
Transfer Function

Temporal **diffusion**
system:

$$\frac{dy(t)}{dt} = c(x(t) - y(t)).$$

- c : diffusion coefficient.

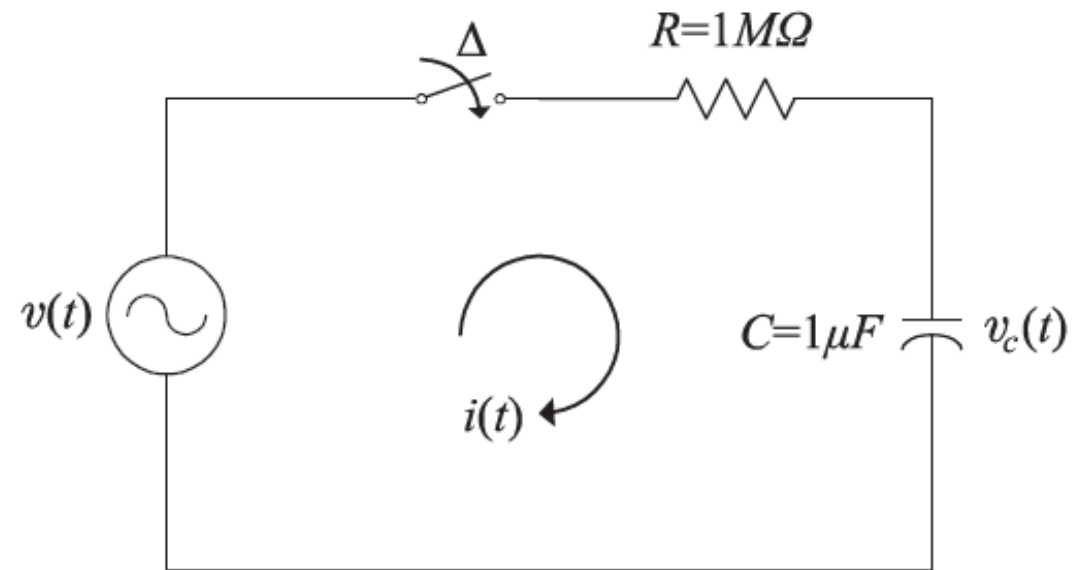
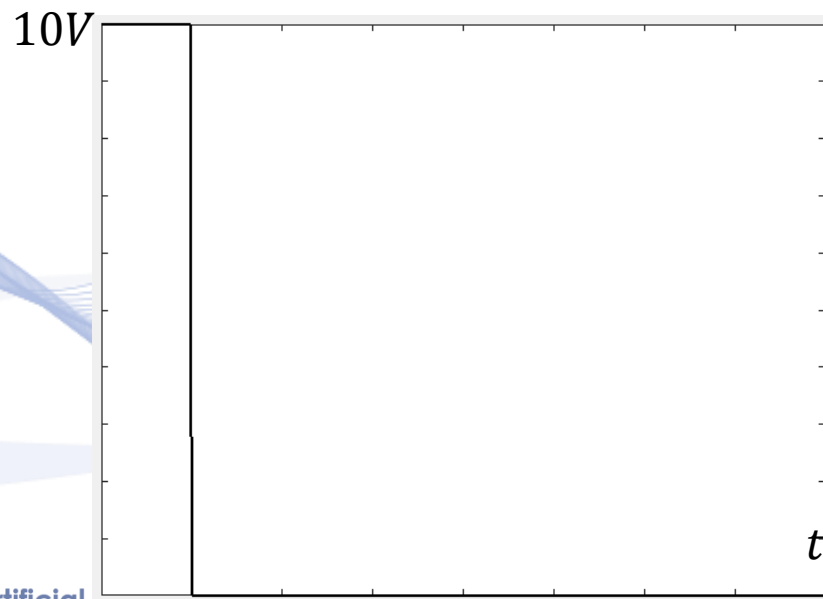
high concentration



low concentration

Transfer Function

RC circuit. The circuit switch turns on at $t = 0$ and the voltage $v(t)$ has the form:

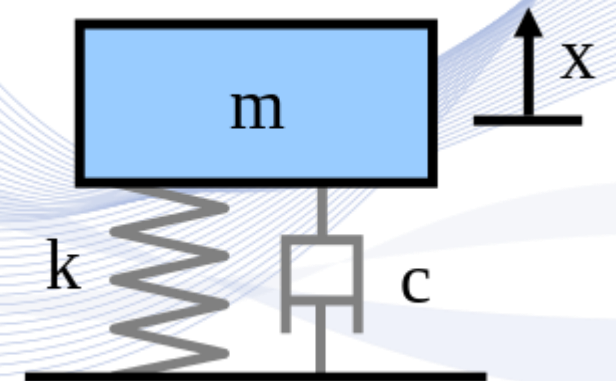


Transfer Function

Mass-spring-damper mechanical system.

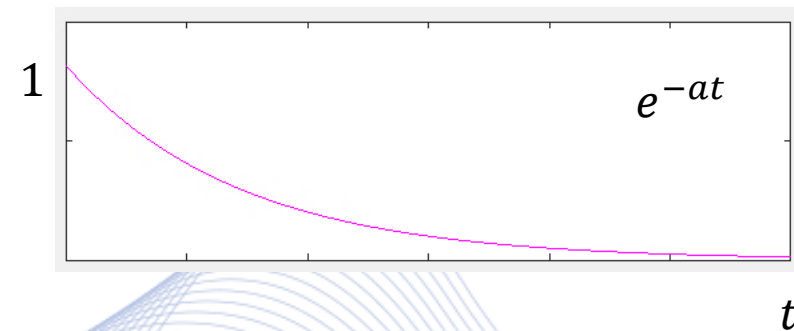
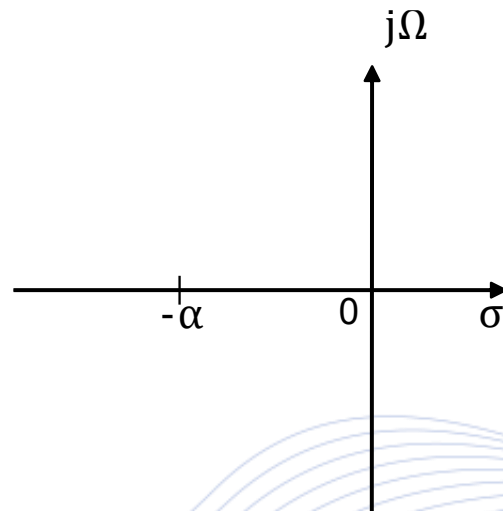
If force $f(t)$ is exercised on a mass m that is attached to a spring having constant k that follows Hooke's law and to a damper having damping constant β , the displacement $y(t)$ is given by the exercised forces:

$$m \frac{d^2 y(t)}{dt^2} + \beta \frac{dy(t)}{dt} + ky(t) = f(t).$$



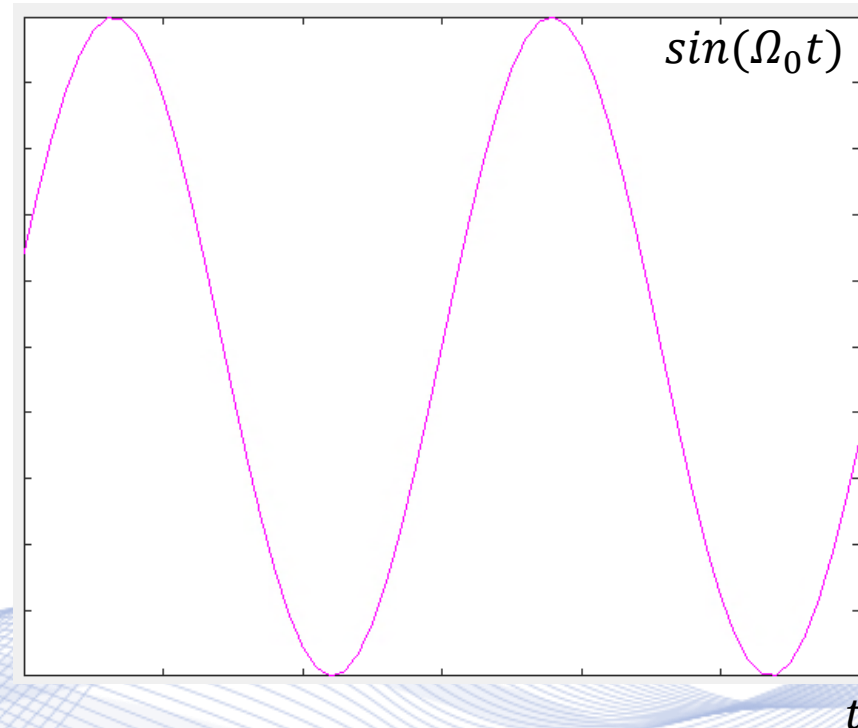
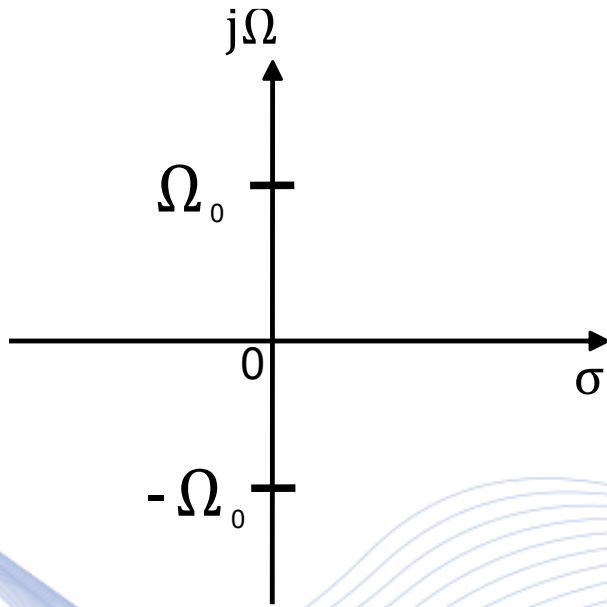
- It models car suspension systems.

System Stability



a) The pole is found in the left complex half-plane, thus the system is stable. b) Impulse response.

System Stability



a) The pole is found on the imaginary axis, thus the system oscillates. b) Impulse response.

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Q & A

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**Contact: Prof. I. Pitas
pitass@csd.auth.gr**